

# Finding a source inside a sphere

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## Abstract

A sphere excited by an interior point source or a point dipole gives a simplified yet realistic model for studying a variety of applications in medical imaging. We suppose that there is an exterior field (transmission problem) and that the total field on the sphere is known. We give analytical inversion algorithms for determining the interior physical characteristics of the sphere as well as the location, strength and orientation of the source/dipole. We start with static problems (Laplace's equation) and then proceed to acoustic problems (Helmholtz equation).

## 1. Introduction

Sixty years ago, Wilson and Bayley [1] began with an assumption 'that the electric field generated by the heart may be regarded as not significantly different from that of a current dipole at the center of a homogeneous spherical conductor', and then went on to examine the effect of moving the dipole away from the center. Point dipoles inside spheres (and ellipsoids) are also used in the context of brain imaging; see Dassios [2] for a recent review and the book by Ammari [3]. Locating point sources and dipoles using surface measurements is an example of an *inverse source problem* [4].

The basic static problem consists of Laplace's equation in a bounded region  $V_i$  with boundary  $\partial V$ . There is a source of some kind in  $V_i$  and the goal is to identify the source from Cauchy data on  $\partial V$ . Uniqueness [5] and stability [6, 7] results are available. For some numerical results, see [8–11] in two dimensions and [12–16] in three dimensions. We will show that some of these problems for balls can be solved exactly by analytical methods. Exact methods for magnetostatic problems have been devised previously [17, 18]; our method is arguably more elementary.

Analogous problems for the Helmholtz equation can be posed [17, 19]. In particular, the problem of low-frequency internal source excitation of a homogeneous sphere has applications in electroencephalography (EEG). The basic principle of EEG lies in the detection and

processing of a signal generated by neural activity in order to map certain brain functions. More precisely, the primary source inside the brain is determined from a set of signals measured on the scalp, thus generating electric brain images [20]. The frequency,  $f$ , of the measured signal from a human brain is very low and hence the interior excitation of a spherical human head by a low-frequency point source constitutes a suitable EEG model. For example [21],  $k_i a \simeq 1.3 \times 10^{-7}$  for  $f = 60$  Hz and head radius  $a = 10$  cm, where  $k_i$  is the interior wavenumber. Further applications arise in magnetic resonance imaging (MRI) [22], brain electrical impedance tomography (EIT) [23], the modeling of antennas implanted inside the head for hyperthermia or biotelemetry [24] and in studies of the operation of wireless devices around sensitive medical equipment (such as cardiac pacemakers) [25].

The direct problem of interior point-source excitation of a layered sphere has been solved in [26] for acoustics and in [27] for electromagnetics. These papers include approximations of the far field in the low-frequency regime and related far-field inverse scattering algorithms. The possibility of using a near-field quantity, namely the scattered field at the source point, in order to obtain inverse scattering algorithms for a small homogeneous soft sphere has been pointed out in [28]; these algorithms are not applicable here as we regard the source point as being inaccessible. For other implementations of near-field inverse problems, see also [29] and [30, p 133].

In this paper, we suppose that there is a point source inside a sphere. There are fields both inside and outside the sphere, with appropriate interface conditions on the sphere. The inverse problem is to determine the location and strength of the source knowing the (total) field on the sphere. For a point dipole, the orientation of the dipole is also to be determined. The internal characteristics (such as wavenumber or conductivity) are also to be found. For static problems, exact and complete results are obtained. Similar results are obtained for acoustic problems, although further approximations are used. The static problem with several point sources is also considered.

Our analytical inversion algorithms make use of the moments obtained by integrating the product of the total field on the spherical interface with spherical harmonic functions. All the information about the primary source and the sphere's interior characteristics is encoded in these moments.

We emphasise that our method is simple, explicit and exact (given exact data). However, as might be expected, it is limited to simple problems (with some extensions mentioned in section 5). We observe that all the inverse problems mentioned above are finite-dimensional, in the sense that the goal is to determine a few numbers (such as the coordinates and strength of the unknown source), not functions. As such, these problems are essentially simpler than many other inverse problems (such as determining the shape of a scatterer), and so it is perhaps not surprising that some of them can be solved analytically. It does not seem to have been noticed that locating a source inside a sphere is one of those problems.

## 2. Mathematical formulation

Consider a homogeneous spherical object of radius  $a$ , surrounded by an infinite homogeneous medium. Denote the exterior by  $V_e$  and the interior by  $V_i$ . There is an interior point source of some kind at an unknown location  $\mathbf{r}_1 \in V_i$ . We are interested in characterizing the source, using information on the spherical interface.

Denote the field outside the sphere by  $u_e$  and the (total) field inside by  $u_i$ . Then, we can write  $u_i = u^{\text{pr}} + u^{\text{sec}}$ , where  $u^{\text{pr}}$  is the primary field due to the source ( $u^{\text{pr}}$  is singular at  $\mathbf{r}_1$ ) and  $u^{\text{sec}}$  is the secondary (regular) field. The field  $u_e$  is regular and satisfies an appropriate far-field

condition. The fields  $u_e$  and  $u_i$  are related by continuity (transmission) conditions across the spherical interface.

Introduce spherical polar coordinates  $(r, \theta, \phi)$  for the point at  $\mathbf{r}$  so that the source is at  $(r_1, \theta_1, \phi_1)$  with  $r_1 = |\mathbf{r}_1| < a$ . Then, the transmission conditions are

$$u_e = u_i \quad \text{and} \quad \frac{1}{\rho_e} \frac{\partial u_e}{\partial r} = \frac{1}{\rho_i} \frac{\partial u_i}{\partial r} \quad \text{at } r = a, \quad (1)$$

where  $\rho_e$  and  $\rho_i$  are constants. To complete the problem formulation, we must specify the governing differential equations, so we separate into electrostatics and acoustics.

### 2.1. Static problems

For static problems, both  $u_e$  and  $u^{\text{sec}}$  are governed by Laplace's equation. The field  $u_e$  decays to zero at infinity. In the context of electrostatics,  $\rho_e$  and  $\rho_i$  are inverse conductivities.

In some applications, the exterior is non-conducting. Then,  $u_i$  solves an interior Neumann problem (with (1) replaced by  $\partial u_i / \partial r = 0$  on  $r = a$ ), and  $u_e$  solves an exterior Dirichlet problem, with  $u_e = u_i$  on  $r = a$ .

For the primary field, we could choose a point source

$$u^{\text{pr}}(\mathbf{r}; \mathbf{r}_1) = \frac{A}{|\mathbf{r} - \mathbf{r}_1|}, \quad \mathbf{r} \in \mathbb{R}^3 \setminus \{\mathbf{r}_1\}, \quad (2)$$

where  $A$  is a real constant; see, for example, [31, p 49]. Dipole fields will also be considered; see (11).

### 2.2. Acoustic problems

For time-harmonic acoustic problems, we consider compressible fluids so that  $\rho_e$  and  $\rho_i$  are densities. The governing equations are Helmholtz equations,

$$(\nabla^2 + k_e^2)u_e = 0 \quad \text{in } V_e, \quad (\nabla^2 + k_i^2)u^{\text{sec}} = 0 \quad \text{in } V_i,$$

where  $k_e$  and  $k_i$  are the respective wavenumbers. The physical fields are given by  $\text{Re} \{u_e e^{-i\omega t}\}$ , for example; henceforth, we suppress the time dependence. The complex-valued field  $u_e$  must satisfy the Sommerfeld radiation condition at infinity. For the primary field, we take

$$u^{\text{pr}}(\mathbf{r}; \mathbf{r}_1) = \frac{A}{|\mathbf{r} - \mathbf{r}_1|} \exp(ik_i|\mathbf{r} - \mathbf{r}_1|), \quad \mathbf{r} \in \mathbb{R}^3 \setminus \{\mathbf{r}_1\}, \quad (3)$$

generated by a point source with the position vector  $\mathbf{r}_1$ , where  $A$  is a (possibly complex) constant; see, for example, [31, p 144] or [32, p 153].

## 3. Static source inside a homogeneous sphere

There is a static source of some kind at  $\mathbf{r}_1$  generating the field  $u^{\text{pr}}$ . Near the sphere ( $r_1 < r < a$ ), separation of variables gives the expansion

$$u^{\text{pr}}(\mathbf{r}; \mathbf{r}_1) = \sum_{n=0}^{\infty} \sum_{m=-n}^n f_n^m(\mathbf{r}_1) (a/r)^{n+1} Y_n^m(\hat{\mathbf{r}}),$$

where  $\hat{\mathbf{r}} = \mathbf{r}/|\mathbf{r}| = (\sin \theta \cos \phi, \sin \theta \sin \phi, \cos \theta)$  and  $Y_n^m(\hat{\mathbf{r}}) = Y_n^m(\theta, \phi)$  is a spherical harmonic; we define  $Y_n^m$  as in [32, section 3.2].

The quantities  $f_n^m$  characterize the source. For a point source, defined by (2),

$$f_n^m(\mathbf{r}_1) = \frac{4\pi A}{a} \frac{(-1)^m}{2n+1} (r_1/a)^n Y_n^{-m}(\hat{\mathbf{r}}_1); \quad (4)$$

see, for example, [33, equation (3.70)].

The total field inside the sphere is  $u_i = u^{\text{pr}} + u^{\text{sec}}$  where

$$u^{\text{sec}}(\mathbf{r}) = \sum_{n=0}^{\infty} \sum_{m=-n}^n \alpha_n f_n^m(\mathbf{r}_1) (r/a)^n Y_n^m(\hat{\mathbf{r}}), \quad 0 \leq r < a$$

( $u^{\text{sec}}$  must be finite at  $r = 0$ ) whereas the field outside is given by

$$u_e(\mathbf{r}) = \sum_{n=0}^{\infty} \sum_{m=-n}^n \beta_n f_n^m(\mathbf{r}_1) (a/r)^{n+1} Y_n^m(\hat{\mathbf{r}}), \quad r > a$$

( $u_e$  must vanish as  $r \rightarrow \infty$ ). The transmission conditions at  $r = a$ , (1), give

$$1 + \alpha_n = \beta_n, \quad -n - 1 + n\alpha_n = -\varrho(n+1)\beta_n,$$

where  $\varrho = \rho_i/\rho_e$ . These can be solved for  $\alpha_n$  and  $\beta_n$  yielding

$$\alpha_n = \frac{(1-\varrho)(n+1)}{n+\varrho(n+1)}, \quad \beta_n = \frac{2n+1}{n+\varrho(n+1)}. \quad (5)$$

Note that these quantities do not depend on any characteristics of the source. Also, for  $\varrho = 1$  (no interface at  $r = a$ ), we can verify that  $u_e(\mathbf{r}) = u^{\text{pr}}(\mathbf{r}; \mathbf{r}_1)$  and  $u^{\text{sec}}(\mathbf{r}) = 0$ , as expected.

The field on the sphere is

$$u_{\text{surf}}(\theta, \phi) = \sum_{n=0}^{\infty} \sum_{m=-n}^n \frac{2n+1}{n+\varrho(n+1)} f_n^m(\mathbf{r}_1) Y_n^m(\theta, \phi).$$

It is this quantity that we will use to find the source.

The spherical harmonics are orthonormal:

$$\int_{\Omega} Y_n^m \overline{Y_n^{\mu}} d\Omega = \int_0^{\pi} \int_{-\pi}^{\pi} Y_n^m(\theta, \phi) \overline{Y_n^{\mu}(\theta, \phi)} \sin\theta d\phi d\theta = \delta_{n\mu} \delta_{m\mu},$$

where  $\Omega$  is the unit sphere and the overbar denotes complex conjugation. Hence, the moments

$$M_n^m \equiv \frac{1}{\sqrt{4\pi}} \int_{\Omega} u_{\text{surf}} \overline{Y_n^m} d\Omega = \frac{1}{\sqrt{4\pi}} \frac{2n+1}{n+\varrho(n+1)} f_n^m(\mathbf{r}_1) \quad (6)$$

are known, in principle, if  $u$  is known on  $r = a$ ; the double integral over  $\Omega$  could be approximated using a suitable quadrature rule and corresponding point evaluations of  $u_{\text{surf}}$ . The problem now is to determine properties of the source and the interior material (namely,  $\rho_i = \rho_e \varrho$ ) from  $M_n^m$ .

### 3.1. Point source

For a point source, (4) and (6) give

$$M_n^m = (-1)^m \frac{\tilde{A} \tilde{r}_1^n \sqrt{4\pi}}{n+\varrho(n+1)} Y_n^{-m}(\theta_1, \phi_1), \quad \text{with } \tilde{A} = \frac{A}{a} \text{ and } \tilde{r}_1 = \frac{r_1}{a}. \quad (7)$$

Thus, there are five unknowns,  $\tilde{A}$ ,  $\varrho$ ,  $\tilde{r}_1$ ,  $\theta_1$  and  $\phi_1$ .

As  $Y_0^0 = (4\pi)^{-1/2}$ , we obtain

$$M_0^0 = \tilde{A}/\varrho.$$

This ratio is all that can be recovered if the source is at the sphere's center ( $r_1 = \tilde{r}_1 = 0$ ). So, let us assume now that  $\tilde{r}_1 \neq 0$ .

For  $n = 1$ , we can use [32, equation (8.28)]

$$Y_1^0 = \sqrt{\frac{3}{4\pi}} \cos \theta, \quad Y_1^1 = -\sqrt{\frac{3}{8\pi}} e^{i\phi} \sin \theta, \quad Y_1^{-1} = \sqrt{\frac{3}{8\pi}} e^{-i\phi} \sin \theta \quad (8)$$

giving

$$M_1^0 = \tilde{A} \frac{\tilde{r}_1 \sqrt{3}}{1 + 2\varrho} \cos \theta_1, \quad M_1^{\pm 1} = \mp \tilde{A} \frac{\tilde{r}_1 \sqrt{3/2}}{1 + 2\varrho} e^{\mp i\phi_1} \sin \theta_1.$$

If  $M_1^{\pm 1} = 0$ , then  $\theta_1 = 0$  or  $\pi$  (the source is on the  $z$ -axis and so  $\phi_1$  is irrelevant); to decide which, note that the sign of  $M_0^0 M_1^0$  is the sign of  $\cos \theta_1$ . If  $M_1^{\pm 1} \neq 0$ ,  $\phi_1$  is determined by noting that the complex number  $M_0^0 M_1^{-1}$  has argument  $\phi_1$ .

If  $M_1^0 = 0$ , then  $\theta_1 = \pi/2$ . If  $M_1^0 \neq 0$ ,  $\sqrt{2} M_1^{-1}/M_1^0 = e^{i\phi_1} \tan \theta_1$  determines  $\theta_1$ . Also

$$(1 + 2\varrho)^2 \{ (M_1^0)^2 - 2M_1^1 M_1^{-1} \} = 3(\tilde{A} \tilde{r}_1)^2. \quad (9)$$

To conclude, we take a measurement with  $n = 2$ . Thus,

$$\varrho(2 + 3\varrho) M_0^0 M_2^m = (\tilde{A} \tilde{r}_1)^2 (-1)^m \sqrt{4\pi} Y_2^{-m}(\theta_1, \phi_1). \quad (10)$$

Then, choosing  $m$  such that  $Y_2^{-m}(\theta_1, \phi_1) \neq 0$  (which means take  $m = 0$  unless  $P_2(\cos \theta_1) = 0$ ), eliminate  $(\tilde{A} \tilde{r}_1)^2$  between (9) and (10) to give a quadratic equation for  $\varrho$  (which is real and positive). Then,  $\tilde{A} = A/a = M_0^0 \varrho$  and  $\tilde{r}_1 = r_1/a$  follows from  $M_1^m$  or from (9).

### 3.2. Point dipole

Let  $\hat{\mathbf{d}} = (\sin \theta_d \cos \phi_d, \sin \theta_d \sin \phi_d, \cos \theta_d)$  be an unknown unit vector, giving the direction of a point dipole at  $\mathbf{r}_1$ . The field generated is given by

$$u_{\mathbf{r}_1}^{\text{pr}}(\mathbf{r}) = \hat{\mathbf{d}} \cdot \text{grad}_1 (A/|\mathbf{r} - \mathbf{r}_1|), \quad (11)$$

where  $\text{grad}_1$  denotes the gradient with respect to  $\mathbf{r}_1$  and  $A$  is a real constant. As changing the sign of  $A$  is equivalent to changing the direction of  $\hat{\mathbf{d}}$ , we can assume that  $A$  is positive.

From (2) and (4), the source coefficients are given by

$$f_n^m(\mathbf{r}_1) = \hat{\mathbf{d}} \cdot \text{grad}_1 \left( \frac{4\pi A}{a} \frac{(-1)^m}{2n+1} (r_1/a)^n Y_n^{-m}(\hat{\mathbf{r}}_1) \right), \quad (12)$$

and then  $M_n^m$  is given by (6).

We have  $f_0^0 = 0$ . Then, with  $n = 1$ , we have

$$\frac{a^2}{\sqrt{4\pi}} f_1^0(\mathbf{r}) = \hat{\mathbf{d}} \cdot \text{grad} \left( \frac{A}{\sqrt{3}} z \right) = \frac{A}{\sqrt{3}} \cos \theta_d$$

and

$$\begin{aligned} \frac{a^2}{\sqrt{4\pi}} f_1^{\pm 1} &= \hat{\mathbf{d}} \cdot \text{grad} \left( \frac{\sqrt{4\pi} A}{3} (-r) Y_1^{\mp 1} \right) \\ &= \hat{\mathbf{d}} \cdot \text{grad} \left( \frac{A}{\sqrt{6}} (iy \mp x) \right) = \mp \frac{A}{\sqrt{6}} e^{\mp i\phi_d} \sin \theta_d. \end{aligned}$$

If  $M_1^{\pm 1} = 0$ , then  $\theta_d = 0$  or  $\pi$ , which we can determine by examining the sign of  $M_1^0$ , recalling that  $A > 0$ . If  $M_1^{\pm 1} \neq 0$ ,  $\phi_d$  is determined by noting that  $M_1^{-1}$  is a complex number with argument  $\phi_d$ .

If  $M_1^0 = 0$ , then  $\theta_d = \pi/2$ . If  $M_1^0 \neq 0$ ,  $\sqrt{2} M_1^{-1}/M_1^0 = e^{i\phi_d} \tan \theta_d$  determines  $\theta_d$ .

Thus, we have determined the orientation of the dipole,  $\hat{\mathbf{d}}$ . Also,

$$(1 + 2\varrho)^2 a^4 \{ (M_1^0)^2 - 2M_1^1 M_1^{-1} \} = 3A^2. \quad (13)$$

To find the dipole's location and strength, we move on to  $n = 2$ . Thus,

$$r^2 Y_2^{\pm 2} = \frac{3\sqrt{5}}{\sqrt{4\pi}\sqrt{4!}}(x \pm iy)^2, \quad r^2 Y_2^{\pm 1} = \mp \frac{3\sqrt{5}}{\sqrt{4\pi}\sqrt{3!}}(x \pm iy)z,$$

$$r^2 Y_2^0 = \frac{\sqrt{5}}{\sqrt{4\pi}} \left( z^2 - \frac{1}{2}(x^2 + y^2) \right).$$

Hence,

$$\begin{aligned} \frac{a^3}{\sqrt{4\pi}} f_2^0(\mathbf{r}_1) &= \hat{\mathbf{d}} \cdot \text{grad}_1 \left( \frac{A}{2\sqrt{5}}(2z_1^2 - x_1^2 - y_1^2) \right) \\ &= \frac{Ar_1}{\sqrt{5}} \{2 \cos \theta_1 \cos \theta_d - \sin \theta_1 \sin \theta_d \cos(\phi_1 - \phi_d)\}, \end{aligned} \quad (14)$$

$$\begin{aligned} \frac{a^3}{\sqrt{4\pi}} f_2^{\pm 1}(\mathbf{r}_1) &= \mp \hat{\mathbf{d}} \cdot \text{grad}_1 \left( \frac{3A}{\sqrt{30}}(x_1 \mp iy_1)z_1 \right) \\ &= \mp \frac{3Ar_1}{\sqrt{30}} \{e^{\mp i\phi_d} \cos \theta_1 \sin \theta_d + e^{\mp i\phi_1} \sin \theta_1 \cos \theta_d\}, \end{aligned} \quad (15)$$

$$\begin{aligned} \frac{a^3}{\sqrt{4\pi}} f_2^{\pm 2}(\mathbf{r}_1) &= \hat{\mathbf{d}} \cdot \text{grad}_1 \left( \frac{3A}{2\sqrt{30}}(x_1 \mp iy_1)^2 \right) \\ &= \frac{3Ar_1}{\sqrt{30}} e^{\mp i(\phi_1 + \phi_d)} \sin \theta_1 \sin \theta_d. \end{aligned} \quad (16)$$

Our first goal is to determine  $\theta_1$  and  $\phi_1$  from  $M_2^m$ . Note that if all the moments  $M_2^m$  vanish, then  $r_1 = 0$  (the dipole is at the sphere's center); henceforth, we assume that  $r_1 > 0$ .

Let us first dispose of special cases. Suppose that  $\theta_d = 0$ , in which case

$$\frac{a^3}{\sqrt{4\pi}} f_2^0(\mathbf{r}_1) = \frac{2Ar_1}{\sqrt{5}} \cos \theta_1, \quad \frac{a^3}{\sqrt{4\pi}} f_2^{\pm 1}(\mathbf{r}_1) = \mp \frac{3Ar_1}{\sqrt{30}} e^{\mp i\phi_1} \sin \theta_1$$

and  $f_2^{\pm 2} = 0$ . If  $M_2^{\pm 1} = 0$ , then  $\theta_1 = 0$  or  $\pi$ , which the sign of  $M_2^0$  determines. If  $M_2^{\pm 1} \neq 0$ ,  $\phi_1$  is the argument of  $M_2^{\pm 1}$ . Then, if  $M_2^0 = 0$ ,  $\theta_1 = \pi/2$ , otherwise  $M_2^{\pm 1}/M_2^0$  determines  $\theta_1$ . A similar analysis succeeds when  $\theta_d = \pi$ .

When  $\theta_d = \pi/2$ ,

$$\frac{a^3}{\sqrt{4\pi}} f_2^{\pm 1}(\mathbf{r}_1) = \mp \frac{3Ar_1}{\sqrt{30}} e^{\mp i\phi_d} \cos \theta_1, \quad \frac{a^3}{\sqrt{4\pi}} f_2^{\pm 2}(\mathbf{r}_1) = \frac{3Ar_1}{\sqrt{30}} e^{\mp i(\phi_1 + \phi_d)} \sin \theta_1.$$

So, if  $M_2^{\pm 2} = 0$ ,  $\theta_1 = 0$  or  $\pi$ , and the sign of  $\mp e^{\pm i\phi_d} M_2^{\pm 1}$  gives the sign of  $\cos \theta_1$ . If  $M_2^{\pm 2} \neq 0$ ,  $e^{-i\phi_d} M_2^{\pm 2}$  has the argument  $\phi_1$ . Then, if  $M_2^{\pm 1} = 0$ ,  $\theta_1 = \pi/2$ , otherwise  $M_2^{\pm 2}/M_2^{\pm 1}$  determines  $\theta_1$ .

Now, for  $\theta_d$  different than  $0$ ,  $\pi/2$  and  $\pi$ , we proceed as follows. If  $M_2^{\pm 2} = 0$ , then  $\theta_1 = 0$  or  $\pi$ , and the sign of  $\mp e^{\pm i\phi_d} M_2^{\pm 1}$  gives the sign of  $\cos \theta_1$ . If  $M_2^{\pm 2} \neq 0$ , the angle  $\phi_1$  is determined from (16) by noting that  $e^{-i\phi_d} M_2^{\pm 2}$  has the argument  $\phi_1$ . Then,  $M_2^1/M_2^2$  gives  $\cot \theta_1$ .

Thus, we have now calculated the angular coordinates of the dipole,  $\theta_1$  and  $\phi_1$ , as well as its orientation,  $\theta_d$  and  $\phi_d$ . It remains to determine  $A$ ,  $r_1$  and  $\varrho$ . Now, from (6) and (12), any non-zero  $M_1^m$  gives  $A/(1 + 2\varrho) = m_1$ , say. Similarly, any non-zero  $M_2^m$  gives  $Ar_1/(2 + 3\varrho) = m_2$ , say. So, in order to extract all three unknowns, we have to move on to  $n = 3$ . (For the point-source problem, discussed in section 3.1, we had non-trivial information from  $n = 0$ , so we did not require measurements with  $n = 3$ .) Then, any non-zero  $M_3^m$  gives  $Ar_1^2/(3 + 4\varrho) = m_3$ , say. Eliminating  $A$  and  $r_1$  gives

$$(1 + 2\varrho)(3 + 4\varrho)m_1m_3 - (2 + 3\varrho)^2m_2^2 = 0,$$

a quadratic for  $\varrho$ . Then,  $m_1$  gives  $A$  and  $m_2$  gives  $r_1$ . Of course, the details of the calculation depend on the choices of  $m$  in  $M_n^m$ . If  $P_3(\cos \theta_1) \neq 0$ , we would select  $M_3^0$  and use (see [32, section 3.4])

$$r^3 Y_3^0 = \frac{\sqrt{7}}{\sqrt{4\pi}} \left( z^3 - \frac{3}{2} z(x^2 + y^2) \right),$$

giving

$$\frac{a^4}{\sqrt{4\pi}} f_3^0(\mathbf{r}_1) = \hat{\mathbf{d}} \cdot \text{grad}_1 \left( \frac{A}{\sqrt{7}} \left( z_1^3 - \frac{3}{2} z_1(x_1^2 + y_1^2) \right) \right) = \frac{3Ar_1^2}{\sqrt{7}} \mathcal{F},$$

where  $\mathcal{F} = -\sin \theta_d \cos \theta_1 \sin \theta_1 \cos(\phi_d - \phi_1) + \cos \theta_d (\cos \theta_1 - \frac{1}{2} \sin^2 \theta_1)$ . Hence,

$$M_3^0 = \frac{\sqrt{4\pi}}{a^4} \frac{3\sqrt{7}}{3 + 4\varrho} A r_1^2 \mathcal{F}. \tag{17}$$

This completes the determination of all the parameters of the problem.

### 3.3. $N$ point sources

Suppose there are  $N$  point sources, located at  $\mathbf{r}_j = (x_j, y_j, z_j)$  with spherical polar coordinates  $(r_j, \theta_j, \phi_j)$ ,  $j = 1, 2, \dots, N$ . Suppose for simplicity that each source has the same strength  $A$ , and that  $\varrho$  is known. Thus, there are  $3N + 1$  unknowns. By linearity and (7), the moments are

$$M_n^m = \frac{A(-1)^m \sqrt{4\pi}}{a^{n+1} \{n + \varrho(n + 1)\}} \sum_{j=1}^N r_j^n Y_n^{-m}(\theta_j, \phi_j).$$

The moment  $M_0^0$  determines  $A$ .

Now,  $r^n Y_n^{-m}(\theta, \phi)$  is a homogeneous polynomial of degree  $n$  in  $x, y$  and  $z$  [30, section 2.3]. For each  $n$ , there are  $2n + 1$  distinct polynomials. So, if we collect all the moments from  $n = 1$  to  $n = N_M$ , we will have  $N_M^2 + 2N_M$  pieces of data from which to recover  $3N$  unknowns. Thus, in principle, we can recover the  $N$  locations by solving a system of polynomial equations: exact solutions will be rare.

For two point sources ( $N = 2$ ), we could use the axisymmetric spherical harmonics ( $m = 0$ ) in order to determine  $r_1, r_2, z_1$  and  $z_2$ ; writing down  $M_n^0$  for  $n = 1, 2, 3$  and  $4$  gives enough equations. For example, ignoring multiplicative constants,  $rY_1^0 = z, M_1^0 = z_1 + z_2, r^2 Y_2^0 = 3z^2 - r^2, M_2^0 = 3z_1^2 + 3z_2^2 - r_1^2 - r_2^2$ , and so on. In this way, we can find a quartic equation for each unknown. We can then use other moments (with  $m \neq 0$ ) to recover  $\phi_1$  and  $\phi_2$ . Evidently, this process becomes more complicated as  $N$  is increased.

## 4. Acoustic source inside a homogeneous sphere

In this section, we outline how some of the static results can be generalized to acoustic problems. We suppose that there is a point source at  $\mathbf{r}_1$  generating the field  $u^{\text{pr}}$  defined by (3). Near the sphere ( $r_1 < r < a$ ), we have the expansion [32, theorem 6.4]

$$u^{\text{pr}}(\mathbf{r}; \mathbf{r}_1) = \sum_{n=0}^{\infty} \sum_{m=-n}^n d_n^m(\mathbf{r}_1) \psi_n^m(k_i, \mathbf{r}), \quad d_n^m = 4\pi i k_i A (-1)^m \hat{\psi}_n^{-m}(k_i, \mathbf{r}_1),$$

where  $\psi_n^m(k, \mathbf{r}) = h_n(kr) Y_n^m(\hat{\mathbf{r}})$ ,  $\hat{\psi}_n^m(k, \mathbf{r}) = j_n(kr) Y_n^m(\hat{\mathbf{r}})$ ,  $j_n$  is a spherical Bessel function and  $h_n$  is a spherical Hankel function of the first kind.

The total field inside the sphere is  $u^{\text{pr}} + u^{\text{sec}}$  where

$$u^{\text{sec}}(\mathbf{r}) = \sum_{n=0}^{\infty} \sum_{m=-n}^n \alpha_n d_n^m(\mathbf{r}_1) \hat{\psi}_n^m(k_i, \mathbf{r}), \quad 0 \leq r < a,$$

whereas the (radiating) field outside is given by

$$u_c(\mathbf{r}) = \sum_{n=0}^{\infty} \sum_{m=-n}^n \beta_n d_n^m(\mathbf{r}_1) \psi_n^m(k_c, \mathbf{r}), \quad r > a.$$

The transmission conditions at  $r = a$  give

$$h_n(\kappa_i) + \alpha_n j_n(\kappa_i) = \beta_n h_n(\kappa_e), \quad h_n'(\kappa_i) + \alpha_n j_n'(\kappa_i) = \beta_n w h_n'(\kappa_e),$$

where  $\kappa_i = k_i a$ ,  $\kappa_e = k_e a$  and  $w = (k_e \rho_i)/(k_i \rho_e)$ . These can be solved for  $\alpha_n$  and  $\beta_n$ . (Note again that these quantities do not depend on properties of the source.) For example,

$$\beta_n = i\kappa_i^{-2} \{w j_n(\kappa_i) h_n'(\kappa_e) - j_n'(\kappa_i) h_n(\kappa_e)\}^{-1}. \quad (18)$$

Then, the field on the sphere is

$$u_{\text{surf}}(\theta, \phi) = \sum_{n=0}^{\infty} \sum_{m=-n}^n \beta_n d_n^m(\mathbf{r}_1) h_n(\kappa_e) Y_n^m(\theta, \phi).$$

Our inverse problem is to determine the location of the source ( $r_1$ ,  $\theta_1$  and  $\phi_1$ ), the strength of the source ( $A$ ) and the interior properties ( $\kappa_i$  and  $\rho_i$ ) from  $u_{\text{surf}}$ . We assume that  $\kappa_e$  and  $\rho_e$  are known.

As the spherical harmonics are orthonormal,

$$\begin{aligned} M_n^m &\equiv \frac{1}{\sqrt{4\pi} h_n(\kappa_e)} \int_{\Omega} u_{\text{surf}} \bar{Y}_n^m d\Omega = \beta_n \frac{d_n^m(\mathbf{r}_1)}{\sqrt{4\pi}} \\ &= i k_i A \beta_n j_n(k_i r_1) \sqrt{4\pi} (-1)^m Y_n^{-m}(\theta_1, \phi_1). \end{aligned}$$

In principle,  $M_n^m$  are known if  $u$  is known on  $r = a$ . Let us write

$$B_n = i k_i A \beta_n j_n(K) \quad \text{with} \quad K = k_i r_1; \quad (19)$$

these quantities do not depend on  $\theta_1$  or  $\phi_1$ .

As  $Y_0^0 = (4\pi)^{-1/2}$ , we obtain  $M_0^0 = B_0$ .

For  $n = 1$ , we can use (8), giving

$$M_1^0 = B_1 \sqrt{3} \cos \theta_1, \quad M_1^{\pm 1} = \mp B_1 \sqrt{3/2} e^{\mp i\phi_1} \sin \theta_1.$$

If all three of these vanish,  $B_1 = 0$ , and we move on to  $n = 2$ .

If  $M_1^{\pm 1} = 0$  but  $M_1^0 \neq 0$ , then  $\theta_1 = 0$  or  $\pi$ ,  $\phi_1$  is irrelevant and  $B_1$  is undetermined.

Suppose now that  $M_1^{\pm 1} \neq 0$ . Write  $B_1 = |B_1| e^{i\delta}$ . Then, as  $\sin \theta_1 > 0$ , the complex number  $M_1^{-1}$  has the argument  $\delta + \phi_1$  and the complex number  $-M_1^1$  has the argument  $\delta - \phi_1$ ; thus, we can determine both  $\delta$  and  $\phi_1$ . For  $|B_1|$ , we can use

$$(M_1^0)^2 - 2M_1^1 M_1^{-1} = 3B_1^2 = 3|B_1|^2 e^{2i\delta}.$$

Finally, use  $M_1^0$  to deduce  $\cos \theta_1$  and hence  $\theta_1$ .

To make further progress, we make further assumptions. They are either that  $k_i$  is known or that  $k_i a$  and  $k_e a$  are small.

#### 4.1. $k_i$ is known

Suppose that  $k_i$  is known. Having already determined  $\phi_1$  and  $\theta_1$ , we proceed to determine  $w$ ,  $r_1$  and  $A$ . By considering  $M_n^0$ , we can obtain values for  $B_n$ . Then, the recurrence relation for spherical Bessel functions gives

$$\frac{1}{K} = \frac{j_{n-1}(K) + j_{n+1}(K)}{(2n+1)j_n(K)} = \frac{B_{n-1}/\beta_{n-1} + B_{n+1}/\beta_{n+1}}{(2n+1)B_n/\beta_n}$$



for each  $n \geq 1$ . Equating two of these gives

$$\frac{B_{n-1}/\beta_{n-1} + B_{n+1}/\beta_{n+1}}{(2n+1)B_n/\beta_n} = \frac{B_n/\beta_n + B_{n+2}/\beta_{n+2}}{(2n+3)B_{n+1}/\beta_{n+1}}.$$

Thus,

$$\frac{2n+3}{2n+1} \left( \frac{B_{n-1}}{\beta_{n-1}} + \frac{B_{n+1}}{\beta_{n+1}} \right) \frac{B_{n+1}}{\beta_{n+1}} = \left( \frac{B_n}{\beta_n} + \frac{B_{n+2}}{\beta_{n+2}} \right) \frac{B_n}{\beta_n}.$$

For each  $n \geq 1$ , this is a quadratic equation for  $w$  because  $\beta_n^{-1}$  is linear in  $w$ ; see (18). This equation has the form

$$\mathcal{A}_n w^2 + \mathcal{B}_n w + \mathcal{C}_n = 0, \quad (20)$$

where

$$\begin{aligned} \mathcal{A}_n &= (2n+3)B_{n-1}B_{n+1}j_{n-1}(\kappa_i)h'_{n-1}(\kappa_e)j_{n+1}(\kappa_i)h'_{n+1}(\kappa_e) \\ &\quad - (2n+1)B_{n+2}B_n j_n(\kappa_i)h'_n(\kappa_e)j_{n+2}(\kappa_i)h'_{n+2}(\kappa_e) \\ &\quad + (2n+3)B_{n+1}^2 [j_{n+1}(\kappa_i)h'_{n+1}(\kappa_e)]^2 - (2n+1)B_n^2 [j_n(\kappa_i)h'_n(\kappa_e)]^2, \\ \mathcal{B}_n &= 2(2n+1)B_n^2 j_n(\kappa_i)h'_n(\kappa_e)j'_n(\kappa_i)h_n(\kappa_e) \\ &\quad - 2(2n+3)B_{n+1}^2 j'_{n+1}(\kappa_i)h_{n+1}(\kappa_e)j_{n+1}(\kappa_i)h'_{n+1}(\kappa_e) \\ &\quad + (2n+1)B_{n+2}B_n [j_n(\kappa_i)h'_n(\kappa_e)j'_{n+2}(\kappa_i)h_{n+2}(\kappa_e) \\ &\quad + j'_n(\kappa_i)h_n(\kappa_e)j_{n+2}(\kappa_i)h'_{n+2}(\kappa_e)] \\ &\quad - (2n+3)B_{n-1}B_{n+1} [j_{n-1}(\kappa_i)h'_{n-1}(\kappa_e)j'_{n+1}(\kappa_i)h_{n+1}(\kappa_e) \\ &\quad + j'_{n-1}(\kappa_i)h_{n-1}(\kappa_e)j_{n+1}(\kappa_i)h'_{n+1}(\kappa_e)], \\ \mathcal{C}_n &= (2n+3)B_{n-1}B_{n+1}j'_{n-1}(\kappa_i)h_{n-1}(\kappa_e)j'_{n+1}(\kappa_i)h_{n+1}(\kappa_e) \\ &\quad - (2n+1)B_{n+2}B_n j'_n(\kappa_i)h_n(\kappa_e)j'_{n+2}(\kappa_i)h_{n+2}(\kappa_e) \\ &\quad + (2n+3)B_{n+1}^2 [j'_{n+1}(\kappa_i)h_{n+1}(\kappa_e)]^2 - (2n+1)B_n^2 [j'_n(\kappa_i)h_n(\kappa_e)]^2. \end{aligned}$$

Although the parameter  $w$  (which gives the interior density  $\rho_i$ ) solves the quadratic equation (20) for each integer  $n$ , the second solution may depend on  $n$ .

The distance  $r_1$  can then be determined using  $\beta_0 B_1 j_0(k_i r_1) = \beta_1 B_0 j_1(k_i r_1)$  (see (19)), while the strength  $A$  follows from the expression for  $B_0$ .

#### 4.2. $k_i a$ and $k_e a$ are small

Suppose that  $k_i$  is unknown. Then analytic progress can be made in the low-frequency zone, using the assumptions  $\kappa_i \ll 1$  and  $\kappa_e \ll 1$ . In that case, the coefficients  $\beta_n$  and  $B_n$  have the following leading-order approximations as  $\kappa_i \rightarrow 0$  and  $\kappa_e \rightarrow 0$ :

$$\beta_n \sim \frac{(2n+1)(k_e/k_i)^{n+1}}{n + \varrho(n+1)}, \quad B_n \sim \frac{iA k_e}{n + \varrho(n+1)} \frac{(k_e r_1)^n}{c_n}, \quad (21)$$

where  $c_n = 1 \cdot 3 \cdots (2n-1)$ , with  $c_0 = 1$ . Then,

$$\frac{3B_2 B_0}{B_1^2} = \frac{(2\varrho+1)^2}{\varrho(3\varrho+2)}$$

determines  $\varrho = \rho_i/\rho_e$ ; as expected, this is equivalent to the static formula, given below (10). Then,  $B_1/B_0 = \varrho k_e r_1 / (2\varrho+1)$  determines  $r_1$  and  $\varrho B_0 = i k_e A$  determines the strength  $A$ .

Evidently, we cannot determine  $k_i$  from the leading-order approximation to  $B_n$ , (21), because  $k_i$  does not appear. Thus, one either should use another kind of measurement, or one could use a higher-order approximation. For example,

$$B_0 \sim iA \frac{k_e}{\varrho} \left( 1 - \frac{\kappa_e^2}{2} + \frac{\kappa_i^2}{6} + \frac{\kappa_i^2}{3\varrho} + \frac{1}{6}(k_i r_1)^2 \right)$$

gives an estimate for  $k_i$ .

## 5. Discussion and conclusions

In this paper, we have considered simple inverse problems, locating sources and dipoles using surface data. These are examples of finite-dimensional inverse problems: the goal is to find the numerical values of a few parameters, such as the coordinates and strength of an interior source.

We have shown that the simplest problems (locating one source or one dipole inside a homogeneous sphere with exact data on the sphere) can be solved exactly.

Analytical solutions raise various questions. The first concerns inexact data. As our formulas are exact, the effect of errors in the data can be studied; for a detailed study of related methods, see [34]. Errors can come from the measurements of  $u_{\text{surf}}$  and from numerical integration over the unit sphere; the latter can be reduced by using accurate quadrature rules. Note that we use integrals of the surface data over the whole sphere: this is a limitation of our method. There are numerical algorithms that use data gathered over a piece of the sphere [16].

Next, we may consider extensions to other geometries, such as a layered sphere or an ellipsoid (motivated by EEG applications), or to elastic media: these extensions are feasible. We could use different data on the sphere, such as Neumann data (but note that, as we have solved a transmission problem, Neumann and Dirichlet data are related by solving an exterior boundary-value problem).

We have also seen (section 3.3) that increasing the number of sources leads to more complicated results. For example, if we want to locate two static sources, we have to determine nine numbers, namely, the coordinates and strength of each source and the interior density. Formally, we can make progress with this problem but we lose the attractive simplicity of the single source/dipole solution. Therefore, given that these multi-source inverse problems are finite-dimensional, it may then be better to abandon the analytical approach and resort to a numerical method.

For another class of problems, we could replace the point source by a spherical inclusion and then generate a primary field at  $r = a$ . The direct problems can be solved exactly, so the inverse problem could be tackled: it will be finite dimensional if the spherical inclusion has constant properties. Problems of this type have been discussed by Bonnetier and Vogelius [35]. We could also consider acoustic problems with a small scatterer embedded in the larger sphere: recall that a small sound-soft object scatters as a source whereas a small sound-hard object scatters as a dipole [32, section 8.2]. We plan to investigate some of these problems in the future.

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